



호남대학교 전자공학과

초고주파 회로 설계 : S-Parameter, Smith-chart

윤 태 순
940-5571 / 010-9925-1974
easteryun@gmail.com



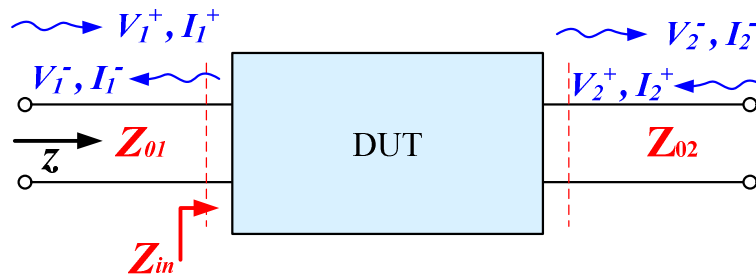
학습 목표

- 회로 특성 파악을 위한 S-parameter 이해
- S-parameter가 갖는 물리적 의미
- 반사계수를 표현한 Smith-chart 이해
- Smith-chart를 통한 임피던스 읽기
- 회로 연결에 따른 Smith-chart에서의 임피던스 이동 경로 파악

easteryun@gmail.com



2-port Network



DUT : device under test

V_1^+ : 1port에서의 입사파 전압, I_1^- : 1port에서의 반사파 전류

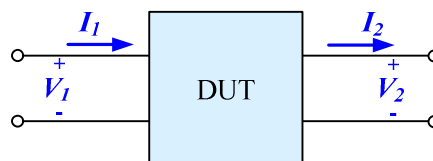
V_2^- : 2port에서의 반사파 전압, I_2^+ : 2port에서의 입사파 전류

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Impedance Matrix



임피던스 파라미터 impedance-parameter

$$\begin{bmatrix} V_1 \\ V_2 \\ \vdots \\ V_N \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} & \cdots & Z_{1N} \\ Z_{21} & \ddots & & \vdots \\ \vdots & & \ddots & \vdots \\ Z_{N1} & \cdots & \cdots & Z_{NN} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ \vdots \\ I_N \end{bmatrix} \quad \text{or} \quad [V] = [Z][I]$$

where Z_{ij} is found from

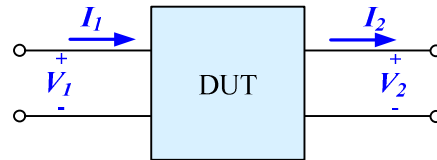
$$Z_{ij} = \left. \frac{V_i}{I_j} \right|_{I_k=0 \text{ for } k \neq j}$$

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Admittance Matrix



어드미턴스 파라미터 admittance-parameter

$$\begin{bmatrix} I_1 \\ I_2 \\ \vdots \\ I_N \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} & \cdots & Y_{1N} \\ Y_{21} & \ddots & & \vdots \\ \vdots & & \ddots & \vdots \\ Y_{N1} & \cdots & \cdots & Y_{NN} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ \vdots \\ V_N \end{bmatrix} \quad \text{or} \quad [I] = [Y][V]$$

where Z_{ij} is found from

$$Y_{ij} = \frac{I_i}{V_j} \bigg|_{V_k=0 \text{ for } k \neq j}$$

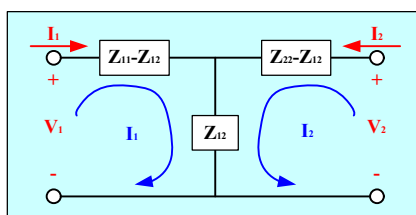
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2-port 회로망의 등가 회로

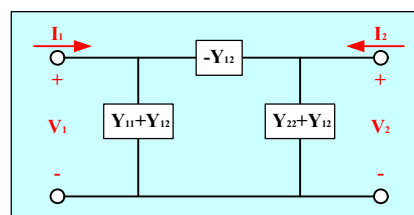
T-equivalent circuit for 2-port network



Kirchhoff voltage law

$$\begin{aligned} V_1 &= I_1(Z_{11} - Z_{12}) + I_1Z_{12} + I_2Z_{12} \\ &= Z_{11}I_1 + Z_{12}I_2 \\ V_2 &= I_2(Z_{22} - Z_{12}) + I_2Z_{12} + I_1Z_{12} \\ &= Z_{12}I_1 + Z_{22}I_2 \end{aligned}$$

pi-equivalent circuit for 2-port network



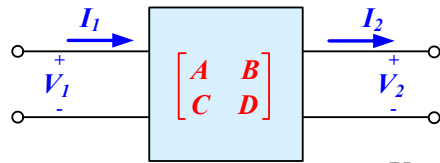
Kirchhoff current law

$$\begin{aligned} I_1 &= V_1(Y_{11} + Y_{12}) - (V_1 - V_2)Y_{12} \\ &= Y_{11}V_1 + Y_{12}V_2 \\ I_2 &= V_2(Y_{22} + Y_{12}) - (V_2 - V_1)Y_{12} \\ &= Y_{12}V_1 + Y_{22}V_2 \end{aligned}$$

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ABCD parameter



$$V_1 = AV_2 + BI_2$$

$$I_1 = CV_2 + DI_2$$

- ABCD 파라미터 ABCD-parameter

$$\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} V_2 \\ I_2 \end{bmatrix}$$

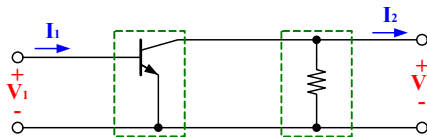
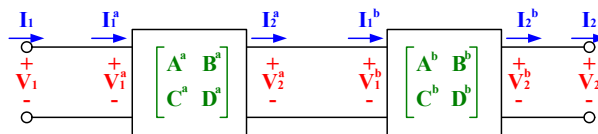
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ABCD-parameter for Cascade Connection

- Cascade connection



$$\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A^a & B^a \\ C^a & D^a \end{bmatrix} \begin{bmatrix} A^b & B^b \\ C^b & D^b \end{bmatrix} \begin{bmatrix} V_2 \\ I_2 \end{bmatrix}$$

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Example of ABCD-parameter

2-port Circuit	ABCD-parameter	
	$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} 1 & Z \\ 0 & 1 \end{bmatrix}$	$V_1 = V_2 + ZI_2$ $I_1 = I_2$
	$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ Y & 1 \end{bmatrix}$	$V_1 = V_2$ $I_1 = YV_2 + I_2$
	$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} \cos \beta l & Z \sin \beta l \\ jY \sin \beta l & \cos \beta l \end{bmatrix}$	$V_1 = V_2 e^{j\beta l} = V_2 \cos \beta l + jV_2 \sin \beta l$ $= V_2 \cos \beta l + jZ_0 I_2 \sin \beta l$ $I_1 = I_2 e^{j\beta l} = jZ_0 V_2 \sin \beta l + I_2 \cos \beta l$
	$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} n & 0 \\ 0 & 1/n \end{bmatrix}$	$V_1 = NV_2$ $I_1 = \frac{1}{N} I_2$

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Example of ABCD-parameter

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} 1 & Z_1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 1/Z_3 & 1 \end{bmatrix} \begin{bmatrix} 1 & Z_2 \\ 0 & 1 \end{bmatrix}$$

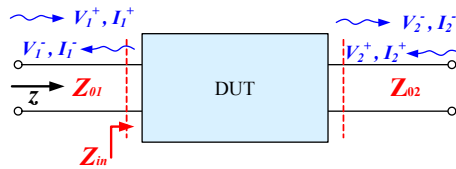
$$= \begin{bmatrix} 1 + \frac{Z_1}{Z_3} & Z_1 + Z_2 + \frac{Z_1 Z_2}{Z_3} \\ \frac{1}{Z_3} & 1 + \frac{Z_2}{Z_3} \end{bmatrix}$$

easteryun@gmail.com

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Scattering Parameter



$$V(x) = V^+(x) + V^-(x)$$

$$I(x) = I^+(x) - I^-(x) = \frac{V^+(x)}{Z_0} - \frac{V^-(x)}{Z_0}$$

$$v(x) = a(x) + b(x) \quad , \quad a(x) = \frac{V^+(x)}{\sqrt{Z_0}} \quad \& \quad b(x) = \frac{V^-(x)}{\sqrt{Z_0}}$$

$$i(x) = a(x) - b(x)$$

$$\begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_N \end{bmatrix} = \begin{bmatrix} S_{11} & S_{12} & \cdots & S_{1N} \\ S_{21} & \ddots & & \vdots \\ \vdots & & \ddots & \vdots \\ S_{N1} & \cdots & \cdots & S_{NN} \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ \vdots \\ a_N \end{bmatrix} \quad \text{or} \quad [b] = [S][a]$$

easteryun@gmail.com

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Scattering Parameter

$$S_{11} = \left. \frac{b_1}{a_1} \right|_{a_2=0}$$

2port가 완전히 termination되어 있을 때의
1port에서의 **입력 반사계수** (reflection coefficient)

$$S_{21} = \left. \frac{b_2}{a_1} \right|_{a_2=0}$$

2port가 완전히 termination되어 있을 때의
1port에서의 **투과계수** (forward transmission coefficient)

$$S_{22} = \left. \frac{b_2}{a_2} \right|_{a_1=0}$$

1port가 완전히 termination되어 있을 때의
2port에서의 **출력 반사계수** (reflection coefficient)

$$S_{12} = \left. \frac{b_1}{a_2} \right|_{a_1=0}$$

1port가 완전히 termination되어 있을 때의
2port에서의 **투과계수** (reverse transmission coefficient)

$$\text{Return Loss} = 10 \log |S_{ii}|^2 \quad [\text{dB}]$$

$$\text{Insertion Loss} = 10 \log |S_{ij}|^2 \quad [\text{dB}]$$

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Scattering Parameter

Reciprocal	[S] matrix is symmetric.
Lossless	[S] matrix is unitary.

- For 2-port network,

$$[S] = \begin{bmatrix} S_{11} & S_{12} \\ S_{21} & S_{22} \end{bmatrix}$$

- For lossless network, it should be $|S_{11}|^2 + |S_{21}|^2 = 1$

- For reciprocal network, it should be $S_{12} = S_{21}$

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Matrix 변환

	S	Z	Y	ABCD
S_{11}		$\frac{(Z_{11} - Z_0)(Z_{22} + Z_0) - Z_{12}Z_{21}}{\Delta Z}$	$\frac{(Y_0 - Y_{11})(Y_0 + Y_{22}) + Y_{12}Y_{21}}{\Delta Y}$	$\frac{A + B/Z_0 - CZ_0 - D}{A + B/Z_0 + CZ_0 + D}$
S_{12}		$\frac{2Z_{12}Z_0}{\Delta Z}$	$\frac{-2Y_{12}Y_0}{\Delta Y}$	$\frac{2(AD - BC)}{A + B/Z_0 - CZ_0 - D}$
S_{21}		$\frac{2Z_{21}Z_0}{\Delta Z}$	$\frac{-2Y_{21}Y_0}{\Delta Y}$	$\frac{2}{A + B/Z_0 - CZ_0 - D}$
S_{22}		$\frac{(Z_{11} + Z_0)(Z_{22} - Z_0) - Z_{12}Z_{21}}{\Delta Z}$	$\frac{(Y_0 + Y_{11})(Y_0 - Y_{22}) + Y_{12}Y_{21}}{\Delta Y}$	$\frac{A + B/Z_0 - CZ_0 - D}{-A + B/Z_0 - CZ_0 + D}$
Z_{11}	$Z_0 \frac{(1 + S_{11})(1 - S_{22}) + S_{12}S_{21}}{(1 - S_{11})(1 - S_{22}) - S_{12}S_{21}}$		$\frac{Y_{22}}{ Y }$	$\frac{A}{C}$
Z_{12}	$Z_0 \frac{2S_{12}}{(1 - S_{11})(1 - S_{22}) - S_{12}S_{21}}$		$\frac{-Y_{12}}{ Y }$	$\frac{AD - BC}{C}$
Z_{21}	$Z_0 \frac{2S_{21}}{(1 - S_{11})(1 - S_{22}) - S_{12}S_{21}}$		$\frac{-Y_{21}}{ Y }$	$\frac{1}{C}$
Z_{22}	$Z_0 \frac{(1 + S_{11})(1 - S_{22}) + S_{12}S_{21}}{(1 - S_{11})(1 - S_{22}) - S_{12}S_{21}}$		$\frac{Y_{11}}{ Y }$	$\frac{D}{C}$
Y_{11}	$Y_0 \frac{(1 - S_{11})(1 + S_{22}) + S_{12}S_{21}}{(1 + S_{11})(1 + S_{22}) - S_{12}S_{21}}$	$\frac{Z_{22}}{ Z }$		$\frac{D}{B}$
Y_{12}	$Y_0 \frac{-2S_{12}}{(1 + S_{11})(1 + S_{22}) - S_{12}S_{21}}$	$\frac{-Z_{12}}{ Z }$		$\frac{BC - AD}{B}$
Y_{21}	$Y_0 \frac{-2S_{21}}{(1 + S_{11})(1 + S_{22}) - S_{12}S_{21}}$	$\frac{-Z_{21}}{ Z }$		$\frac{-1}{B}$
Y_{22}	$Y_0 \frac{(1 + S_{11})(1 - S_{22}) + S_{12}S_{21}}{(1 + S_{11})(1 + S_{22}) - S_{12}S_{21}}$	$\frac{Z_{11}}{ Z }$		$\frac{A}{B}$
A	$\frac{(1 + S_{11})(1 - S_{22}) + S_{12}S_{21}}{2S_{21}}$	$\frac{Z_{11}}{ Z_{21} }$	$\frac{-Y_{22}}{Y_{21}}$	
B	$Z_0 \frac{(1 + S_{11})(1 - S_{22}) + S_{12}S_{21}}{2S_{21}}$	$\frac{ Z }{Z_{12}}$	$\frac{-1}{Y_{21}}$	
C	$\frac{1}{Z_0} \frac{(1 - S_{11})(1 - S_{22}) - S_{12}S_{21}}{2S_{21}}$	$\frac{1}{Z_{21}}$	$\frac{- Y }{Y_{21}}$	
D	$\frac{(1 - S_{11})(1 + S_{22}) + S_{12}S_{21}}{2S_{21}}$	$\frac{Z_{22}}{Z_{21}}$	$\frac{-Y_{11}}{Y_{21}}$	



스미스 차트 (Smith Chart)

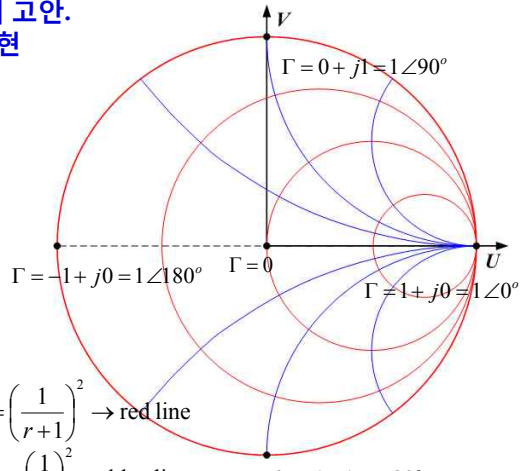
- 1939년에 Phillip Smith에 의해 고안.
- 복소수 평면에서 반사계수 표현

$$\Gamma = \frac{z_L - 1}{z_L + 1} \quad \leftarrow \quad z_L = r + jx$$

$$= \frac{(r-1) + jx}{(r+1) + jx} = U + jV$$

$$U = \frac{r^2 - 1 + x^2}{(r+1)^2 + x^2} \Rightarrow \left(U - \frac{r}{r+1} \right)^2 + V^2 = \left(\frac{1}{r+1} \right)^2 \rightarrow \text{red line}$$

$$V = \frac{2x}{(r+1)^2 + x^2} \Rightarrow (U-1)^2 + \left(V - \frac{1}{2} \right)^2 = \left(\frac{1}{x} \right)^2 \rightarrow \text{blue line}$$

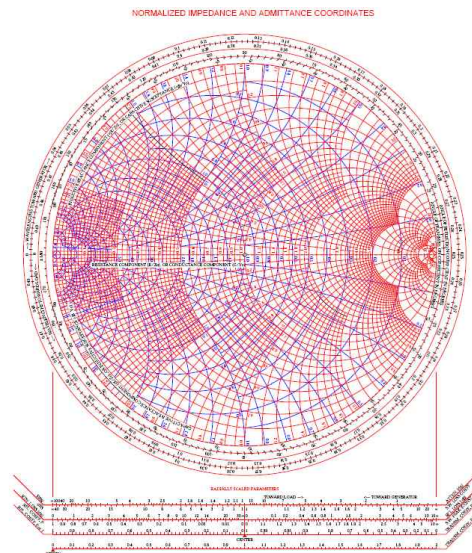


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스미스 차트 (Smith Chart)

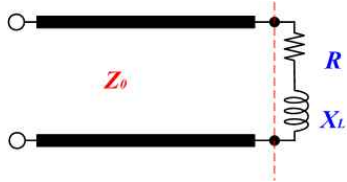


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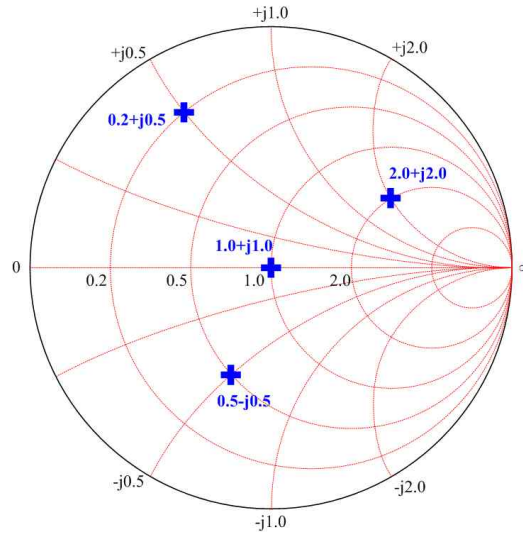
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반사 계수의 표현



$$\Gamma = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{z_L - 1}{z_L + 1}$$



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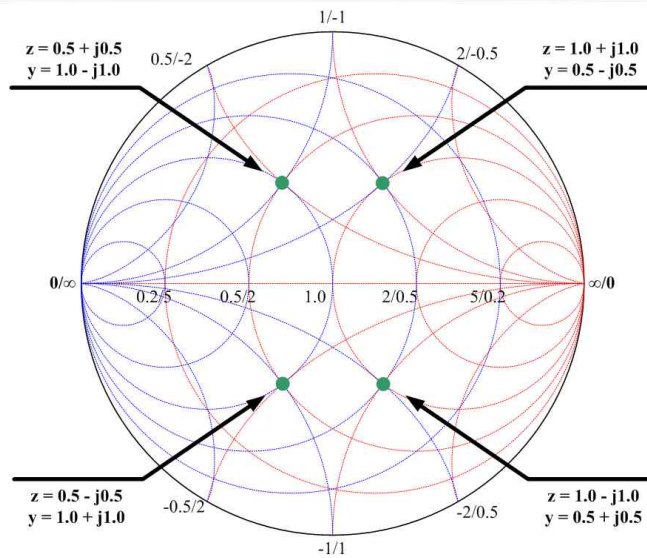


임피던스 & 어드미턴스 차트

$$\begin{aligned} \Gamma_0 &= \frac{Z_L - Z_0}{Z_L + Z_0} \\ &= \frac{z_L - 1}{z_L + 1} \\ &= \frac{y_0 - y_L}{y_0 + y_L} \end{aligned}$$

$$z = \frac{Z}{Z_0}$$

$$y = \frac{Y}{Y_0}$$

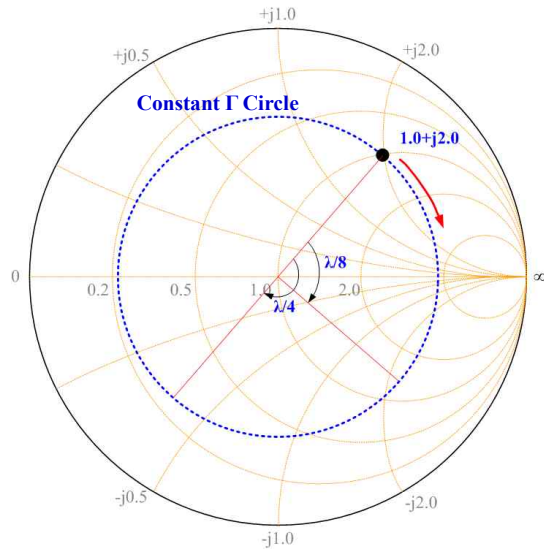
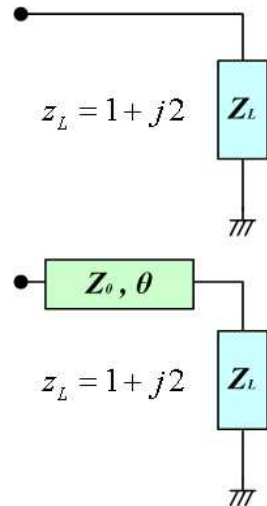


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전송선로 연결에 따른 반사 계수 변화

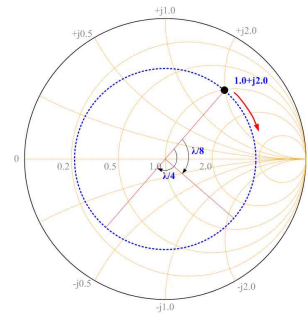
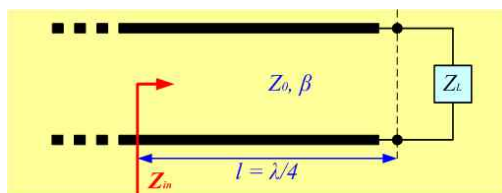


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$\lambda/4$ 전송선로 연결



$$Z_{in} = Z_0 \frac{Z_L + jZ_0 \tan \beta l}{Z_0 + jZ_L \tan \beta l} \bigg|_{l=\lambda/4} = \frac{Z_0^2}{Z_L}$$

$$z_{in} \bigg|_{l=\lambda/4} = \frac{1}{z_L} = y_L$$

open circuit $\longleftrightarrow$ short circuit
 $\lambda/4$

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Example : Connection of Transmission-line



S-PARAMETERS

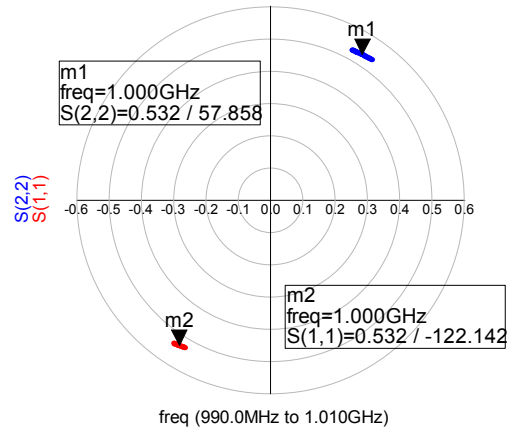
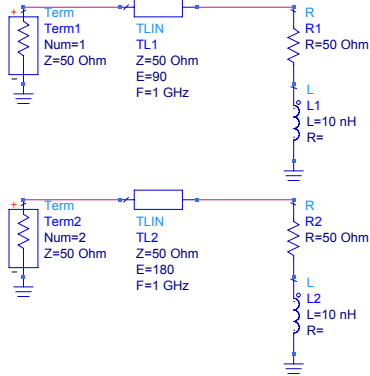
S_Param

SP1

Start=0.99 GHz

Stop=1.01 GHz

Step=0.01 MHz



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LC 연결에 따른 반사 계수 변화

Series Inductor (SL)



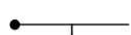
$$z = \frac{j\omega L}{Z_0}$$

Series Capacitor (SC)



$$z = \frac{-j}{\omega C Z_0}$$

Parallel Inductor (PL)

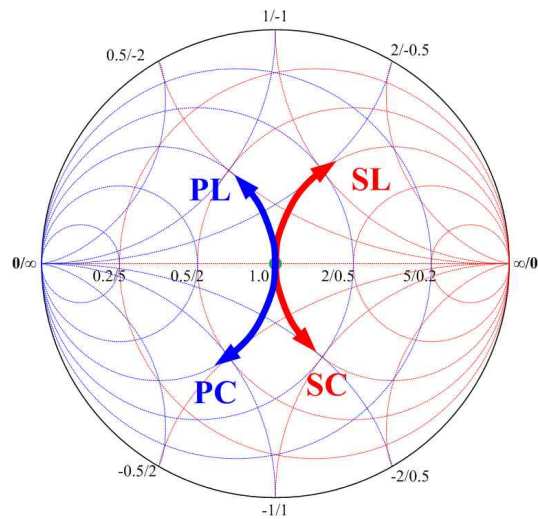


$$z = \frac{-j}{\omega L Y_0}$$

Parallel Capacitor (PC)



$$z = \frac{j\omega C}{Y_0}$$

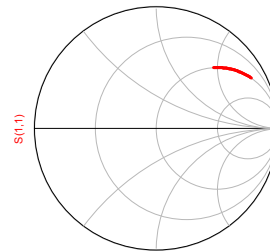
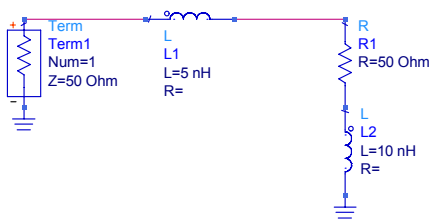
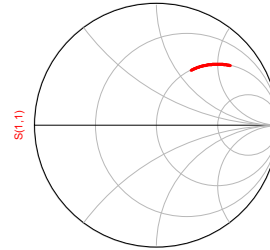
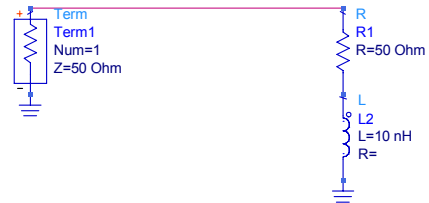


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Example : Connection of Series L

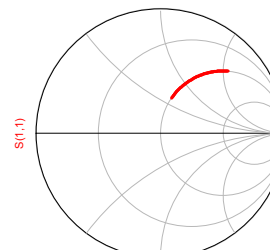
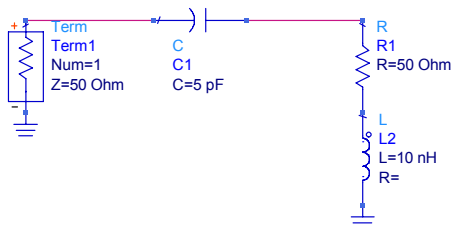
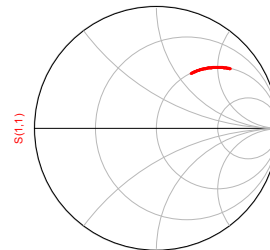
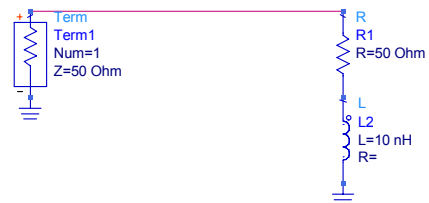


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Example : Connection of Series C

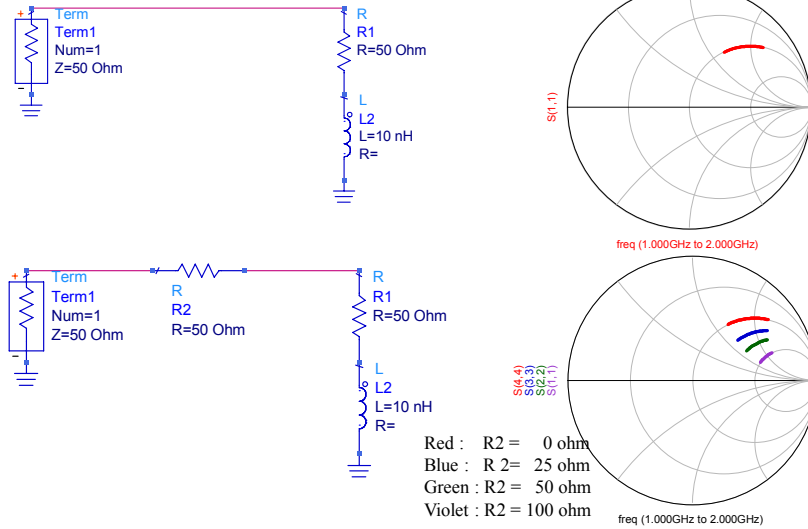


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Example : Connection of Series R

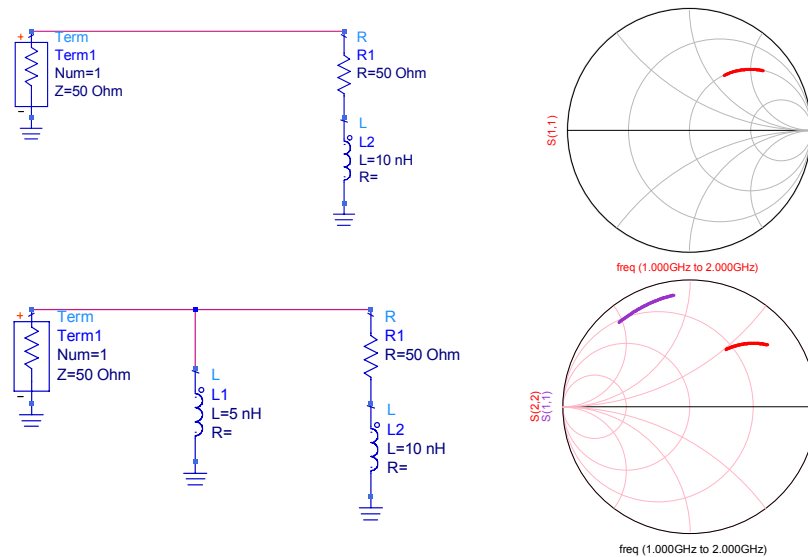


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Example : Connection of Parallel L

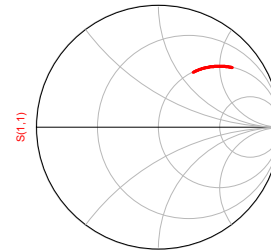
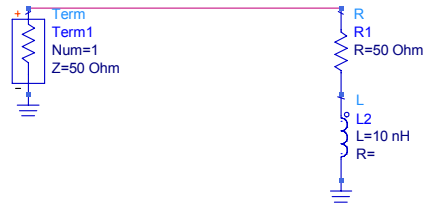


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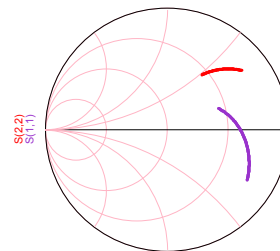
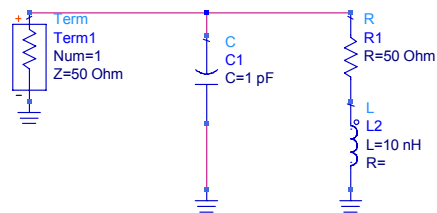
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Example : Connection of Parallel C



freq (1.000GHz to 2.000GHz)



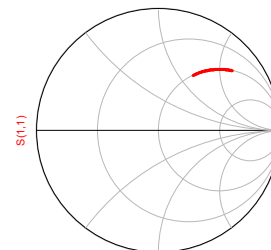
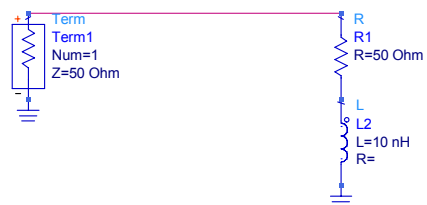
freq (1.000GHz to 2.000GHz)

easteryun@gmail.com

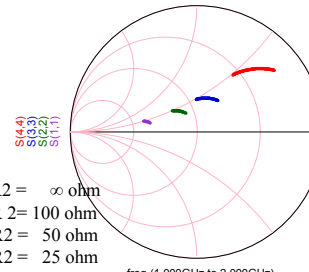
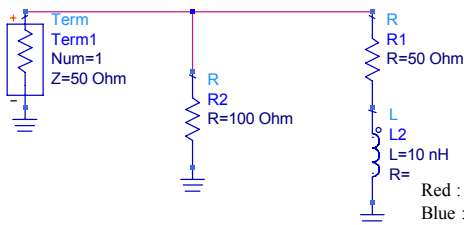
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Example : Connection of Parallel R



freq (1.000GHz to 2.000GHz)



freq (1.000GHz to 2.000GHz)

Red : $R2 = \infty$ ohm
Blue : $R2 = 100$ ohm
Green : $R2 = 50$ ohm
Violet : $R2 = 25$ ohm

easteryun@gmail.com

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